

Statistical tools for anomaly detection as a part of predictive maintenance in the mining industry

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We present new achievements in the area of anomaly detection related to predictive maintenance in the mining industry. The main focus is on the problem of local damage detection based on vibration signals analysis. The vibration signals acquired from machines usually have a complex spectral structure. As the signal of interest (SOI) is weak (especially at an early stage of damage) and covers some frequency range, it must be extracted from raw observations. Up to now, most the techniques assumed the presence of Gaussian noise. However, there are cases in which the non-informative part of the signal (considered as the noise) is non-Gaussian due to random disturbances or to the nature of the process executed by the machine. In such cases, the problem can be formulated as the extraction of the SOI from the non-Gaussian noise. Recently, the importance of this problem has been recognised by several authors, and some new ideas have been developed. We present here a comparison of the new techniques for benchmark signals. Our analysis will cover classical approaches and recently introduced algorithms based on the stochastic analysis of the vibration signals with non-Gaussian distribution.

1 Introduction

Vibration-based condition monitoring is commonly used for the maintenance of mechanical systems [35, 36]. The main focus is usually set on gearboxes and bearings, as these elements appear in most transmission systems and their failure is the most frequent reason for a machine breakdown; for recent reviews, see [16, 42].

From the mathematical/statistical point of view, the task is defined as the detection of the periodic impulsive behaviour in the vibration signal. The most popular approach is the envelope analysis (and its various modifications) and detection of fault frequencies in the spectrum of the envelope for a given signal; see e.g. [5, 6, 37, 40]. The filtration of a raw signal is used to select its informative part and avoid other spectral content not related to the local damage.

The most popular approach is based on spectral kurtosis as an informative frequency band (IFB) selector (filter characteristic). The kurtosis value is calculated for sub-signal at some narrow frequency band. As kurtosis is sensitive to outliers [41], one can

select impulsive content at a given narrow frequency band and filter out other components. Kurtosis is the most intuitive statistic commonly used for machine diagnostics [36]. It has plenty of variations and extensions, e.g. the kurtogram [1] which is a coloured map in which the depth of the colour values is proportional to the kurtosis value. Moreover, other statistics are also used in such context [33].

In the literature, one can also find methods which are based on the cyclostationary approach. Recall that cyclostationary signals are considered when some of their characteristics are periodic in time. The most common characteristic used in this context is the autocovariance function, in which case we consider the cyclostationarity of the second order. The methods for the analysis of cyclostationary signals are dedicated to the cyclic behaviour identification. In the classical approach of cyclostationary-based techniques, the Gaussian distribution of the signal is usually assumed [2, 6, 7, 11, 29, 31].

One can also find other approaches for the informative frequency band selection based on artificial intelligence methods [26, 38, 50]. However, there is still a need for new approaches that would allow us to consistently handle restrictions linked to the amount of available data, specific type of noise, work specifications of the tested machine, etc.

Unfortunately, most of the standard statistical indicators used for local damage detection might be not appropriate in the presence of non-Gaussian noise. In the real environment, we observe signals with cyclic and impulsive behaviour. One of the examples is the crushing machine [47] used in mines. During the operation of such a machine (the crushing process), apart from the background noise (which is often assumed to be Gaussian white noise) in the vibration signal, large observations appear due to the nature of the machine's work. Moreover, in the case of local damage, the additional cyclic impulses are hidden in the signal. In this case, the detection of local damage is very difficult. The non-Gaussian noise could be also linked to other technological processes of the working machine and may correspond to milling, sieving, cutting, compressing, etc. In the literature, algorithms for signals with non-Gaussian distribution have been proposed [10, 19, 21, 27, 28, 48, 49]. A new definition of the cyclostationary non-Gaussian signal was introduced in [25].

We demonstrate here some recently proposed algorithms for local damage detection that take into consideration the possible non-Gaussian behaviour of the signal and utilise the advanced statistics that are robust for large observations not related to failure. We present two approaches. The first is related to new techniques that can replace the classical measure of impulsiveness, i.e. kurtosis used as the selector for the IFB. In the second approach, the local damage is detected using the cyclostationary analysis dedicated to non-Gaussian distributed signals. All these techniques were recently published and used for simulated and real signals [12–14, 25, 32]. Methods dedicated to non-Gaussian vibration signals were also developed during the OPMO project (EIT Raw Materials), which was carried out at the Wrocław University of Science and Technology (2019–2021), among others with a global mining industry leader KGHM. Those results were also the basis for two research projects currently implemented at the Faculty of Pure and Applied Mathematics at the Wrocław University of Science and Technology (the first in cooperation with the AMC Tech company and the second with the Tsinghua University, China).

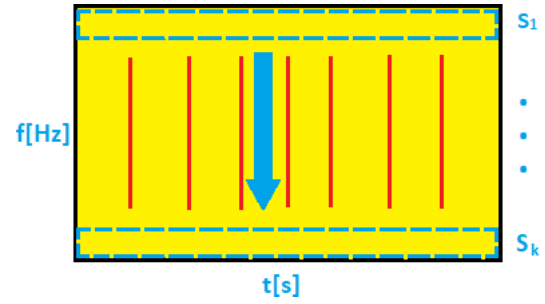
2 Informative frequency band selectors for non-Gaussian signals

In the problem of IFB selection, the first step is the decomposition of the raw signal into a set of narrow-band sub-signals using a time-frequency representation to obtain several dozen time series. To perform signal decomposition, one may use various techniques (wavelets, Wigner, EMD, etc.); see e.g. [9]. In this research, the Short-Time Fourier Transform (STFT) is used. The transform is defined as

$$\text{STFT}(t, f) = \sum_{k=1}^N x_k w(t - k) e^{\frac{-2\pi i f k}{N}},$$

where $w(t - k)$ is the shift window, $x_1, x_2, \dots, x_N$ is the input signal, N is its length, $t \in T$ is the time point, and $f \in F$ is a frequency; see [3] for more details. Interpretation of the STFT is intuitive: the squared envelope of the STFT (spectrogram) describes the energy flow in time for some narrow frequency band, i.e. sub-signal. The simplified representation of the spectrogram is presented in Figure 1 (a).

The next step is the application of some statistics (called selectors) to time series obtained from the spectrogram $S_1, S_2, \dots, S_k$ (Figure 1 (b)) to identify if a given sub-signal fulfils expectations regarding information about faults and select similar sub-signals from the whole frequency range. Distribution of selectors along frequencies, after normalisation, may constitute the filter characteristic (frequency response). One should expect that IFB for the healthy machine will contain stationary noise. In case of damage, the energy flow in some frequency bins will reveal a non-stationary character, and distribution of such sub-signals will be far from Gaussian; see [34, 46, 47] for more details. In Figure 1, we present



(a)

$$\begin{aligned} S_1 &= \{|\text{STFT}(t_1, f_1)|, |\text{STFT}(t_2, f_1)|, \dots, |\text{STFT}(t_i, f_1)|\} \\ S_2 &= \{|\text{STFT}(t_1, f_2)|, |\text{STFT}(t_2, f_2)|, \dots, |\text{STFT}(t_i, f_2)|\} \\ &\vdots \\ S_k &= \{|\text{STFT}(t_1, f_k)|, |\text{STFT}(t_2, f_k)|, \dots, |\text{STFT}(t_i, f_k)|\} \end{aligned}$$

(b)

Figure 1. (a) Simplified representation of the spectrogram, (b) matrix representation of the spectrogram [32]

the general idea of preparing the signal to calculate the selectors for IFB using the time-frequency representation of the signal.

In the following subsections, we present three statistics that are considered in the literature as the IFB selectors.

2.1 Kurtosis

In probability theory and statistics, kurtosis is considered as the tail measure of the probability distribution [43]. For a given random variable X with finite fourth moment, its kurtosis is defined as

$$K = \frac{\mu_4}{\sigma^4},$$

where $\mu_4 = E(X - \mu)^4$ is the fourth central moment of X , μ is its expectation of X , and σ^2 is the variance of X . For a Gaussian random variable, the kurtosis is always equal to 3; sometimes the term excess kurtosis is used in reference to $K - 3$. The empirical kurtosis is based on a scaled version of the fourth empirical moment of the data. Given a signal $x_1, x_2, \dots, x_N$, the empirical kurtosis is a statistic defined as

$$\hat{K} = \frac{\sum_{i=1}^N (x_i - \bar{x})^4}{N \hat{\sigma}^4},$$

where $\hat{\sigma}^2$ is the empirical sample variance and $\bar{x}$ is the empirical mean; see [20] for details. In our case, the statistic $\hat{K}$ is calculated for the time series $S_1, S_2, \dots, S_k$ from time-frequency representation of the raw signal, see Figure 1 (b), and thus in this case it is called the spectral kurtosis [13].

2.2 Alpha selector

Let us first recall the class of α -stable distributions. It is considered as a natural extension (via the generalised limit theorem) of the classical Gaussian distribution; see [15, 39]. It is very useful in data modelling with impulsive behaviour (also for damage detection), since in general it contains heavy-tailed (power-law) distributions. In the problem under consideration, we do not use the α -stable distribution to describe the signal, but to analyse one of the parameters of this distribution (the stability index α) as the measure of impulsiveness, and consider its estimator as the selector for IFB.

The random variable X has a stable distribution with stability index $\alpha \in (0, 2]$, scale parameter $\sigma > 0$, skewness parameter $\beta \in [-1, 1]$, and shift parameter $\mu \in \mathbb{R}$, if its characteristic function $\phi_X(\theta) = \mathbb{E}[\exp\{i\theta X\}]$ is given by

$$\phi_X(\theta) = \begin{cases} e^{-\sigma^\alpha |\theta|^\alpha \{1 - i\beta \operatorname{sign}(\theta) \tan(\pi\alpha/2)\} + i\mu\theta}, & \alpha \neq 1, \\ e^{-\sigma|\theta| \{1 + i\beta \operatorname{sign}(\theta) 2/\pi \log(|\theta|)\} + i\mu\theta}, & \alpha = 1. \end{cases}$$

For $\beta = \mu = 0$, X has a symmetric stable distribution, and for $\alpha = 2$, the α -stable distribution reduces to the Gaussian.

In the literature, one can find different α estimation methods; see e.g. [4, 24]. We use the classical McCulloch method [30]. In further analysis, the statistic $1 - \hat{\alpha}$ obtained using this approach applied to the sub-signals $S_1, S_2, \dots, S_k$ (Figure 1 (b)) is called the Alpha selector; see [13].

2.3 Conditional variance-based selector

The selector called conditional variance-based (CVB) was introduced in [13]. Below, we briefly describe this approach. Let us assume that X is a Gaussian random variable with mean μ and standard deviation σ . Let $\Phi_{\mu,\sigma}(\cdot)$ denote the distribution function (cdf) of X . For any level $0 < q < 0.5$, we define the left, right, and middle quantile partitioning of X by

$$\begin{aligned} L_q &:= (-\infty, \Phi_{\mu,\sigma}^{-1}(q)], \\ R_q &:= [\Phi_{\mu,\sigma}^{-1}(1-q), \infty), \\ M_q &:= (\Phi_{\mu,\sigma}^{-1}(q), \Phi_{\mu,\sigma}^{-1}(1-q)), \end{aligned}$$

where $\Phi_{\mu,\sigma}^{-1}$ is the inverse of $\Phi_{\mu,\sigma}$, i.e. $\Phi_{\mu,\sigma}^{-1}(d)$ denotes the d -quantile of X . Under the normality assumption and the partitioning ratio close to 20/60/20, i.e. for $q \approx 0.2$, we get that

$$\sigma_{L_q}^2 = \sigma_{M_q}^2 = \sigma_{R_q}^2, \quad (1)$$

where $\sigma_A^2 := \operatorname{Var}(X | X \in A)$ is the conditional variance of X on set A ; see [17] for details. Moreover, the 20/60/20 ratio is the unique quantile (three set) partitioning satisfying property (1).

This property can be described as follows: if we split the large normal random sample into three sets, one corresponding to the worst (smallest) 20 % of outcomes, one corresponding to the middle 60 % of outcomes, and one corresponding to the

best (largest) 20 % of outcomes, then the conditional variance on appropriate subsets is approximately the same.

As noted in [17], condition (1) creates a dispersion balance for the conditional populations. This might be linked to the statistical phenomenon commonly referred to as the 20/60/20 rule.

Since the ratio 20/60/20 is the unique ratio for which condition (1) is satisfied, it can be used to construct a goodness-of-fit test statistic. In other words, by comparing conditional variances with the conditional central variance, one can verify whether the sample comes from a Gaussian distribution. As the conditional tail variance might be seen as a measure of a tail heaviness, we can also use property (1) to benchmark any distribution tails with respect to normal tails without making any explicit assumptions about the distribution of X .

The statistic used in [18] for Gaussian distribution testing is defined as

$$\hat{C} := \frac{1}{\rho} \left(\frac{\hat{\sigma}_{L_q}^2 - \hat{\sigma}_{M_q}^2}{\hat{\sigma}^2} + \frac{\hat{\sigma}_{R_q}^2 - \hat{\sigma}_{M_q}^2}{\hat{\sigma}^2} \right) \sqrt{N}, \quad (2)$$

where $q = 0.2$, $\rho \in \mathbb{R}$ is a normalisation constant, $\hat{\sigma}^2$ is the sample variance, and $\hat{\sigma}_A^2$ is set A conditional sample variance. Assuming that the sample is independent and identically distributed, the asymptotic distribution of $\hat{C}$ is standard normal. Moreover, if the sample under consideration comes from a (symmetric) heavy-tailed distribution, the values of the statistic C should be positive due to high values of conditional tail variances on sets L_q and R_q . Consequently, $\hat{C}$ could be considered as the measure of tail fatness, i.e. the bigger the value of $\hat{C}$, the fatter the tails. The value of $\hat{C}$ for real signals is based on the empirical conditional variance. See [13] for more details.

The statistic given in (2) can be extended, and a different number of partitioning sets could be considered. In [13], it was proposed to partition into seven quantile conditioning subsets and use the (unique) ratio guaranteeing conditional variance equality. The statistic used as the CVB selector is defined as

$$\hat{C}_1 := \left(\frac{\hat{\sigma}_{A_3}^2 - \hat{\sigma}_{A_4}^2}{\hat{\sigma}^2} + \frac{\hat{\sigma}_{A_5}^2 - \hat{\sigma}_{A_4}^2}{\hat{\sigma}^2} \right)^2 \sqrt{N},$$

where $\hat{\sigma}_A^2$ is the conditional sample variance on A and the subsets A_i come from the partitioning of the signal into seven appropriate subsets. In this approach, C_1 was used to measure the impact of the non-extreme (trimmed) tail variance on the central part of the distribution without bench-marking the model using Gaussian distribution. Similarly to what we saw above for kurtosis and the Alpha selector, $\hat{C}_1$ applied to the time series $S_1, S_2, \dots, S_k$ (Figure 1 (b)) is used as the selector for IFB.

3 Comparative study of IFB selection – simulated signal analysis

In order to verify the efficiency of the presented selectors, we simulated four different types of signals: s_1 , s_2 , s_3 and s_4 . The first signal s_1 is the Gaussian white noise, which corresponds to the bearing vibration in a healthy condition; see Figure 6 (a). In such a case, we expect any of the informative frequency band selectors to respond significantly. The signal's frequency is 25000 Hz and its length is 1 second. Signal s_2 corresponds to the locally damaged bearing vibration and is defined as

$$s_2 = \text{ACI} \cdot \text{gauspuls}(t_0, fc_0, bw_0) + s_1,$$

where $\text{gauspuls}(t, fc, bw)$ is a unity-amplitude Gaussian radio-frequency (RF) pulse at the times indicated in array t , with a centre frequency fc in hertz and a fractional bandwidth bw . ACI is the amplitude of the cyclic impulses ($\text{ACI} = 3$), fc_0 is set to 2500 Hz, and the frequency modulation of cyclic impulses is equal to 30 Hz. The simulated signal s_2 is presented in Figure 6 (b).

The signal s_3 imitates the bearing operation of the loaded machine in a healthy condition; it is defined as

$$s_3 = \text{ANCI} \cdot \text{gauspuls}(t_1, fc_1, bw_1),$$

where $\text{ANCI} = 30$ is the amplitude of the non-cyclic impulses and t_1 is the location of the non-cyclic impulses, with the uniform distribution. The simulated signal s_3 is depicted in Figure 6 (c).

The last of the signals is a mixture of the previous ones, namely, it is defined as $s_4 = s_2 + s_3$. It imitates the bearing vibrations in the case of a loaded machine operating in the unhealthy condition of bearing; see Figure 6 (d). In Figure 7, we present the spectrograms for the simulated signals presented in Figure 6. Panel (a) illustrates the spectrogram of the signal s_1 . The signal is the Gaussian white noise, so it is neither impulsive nor periodic. In panel (b), we demonstrate the spectrogram for signal s_2 . For this cyclic impulsive signal, we expect the techniques to point out the informative frequency band between 2 and 3 kHz (the centre frequency is set to 2500 Hz). In panel (c), we show the spectrogram for signal s_3 . For this non-cyclic impulsive signal, any IFB is expected. Finally, in panel (d), the spectrogram for signal s_4 is demonstrated. This is the most complicated case. The signal contains large non-cyclic and cyclic impulses. Thus, we expect to find information about the cyclic impulses with possibly suppressed information about non-cyclic ones.

In Figures 8–10, we present the considered selectors for the four simulated signals s_1, s_2, s_3, s_4 . As one can see, if the signal contains only the Gaussian noise (s_1), all considered techniques, kurtosis, Alpha selectors and CVB, have relatively small amplitudes and do not indicate the informative frequency band, as expected. There is no frequency band which significantly stands out from the others. For the signal s_2 , all techniques work well, but the results for the CVB selector seem to be the most unequivocal. The value of the CVB selector in the range of the IFB is significantly higher than for

other frequency bins. If the cyclic impulses have higher amplitudes, then their variance will increase and the value of the CVB will increase as well. For the non-cyclic impulsive signal s_3 , all techniques properly select the frequency band where the non-cyclic impulses appear. Note that none of the methods take into consideration the cyclic behaviour of the signal, only its impulsiveness. For the most complicated case, i.e. for the signal with a large non-cyclic to cyclic impulse amplitude ratio (s_4), surprisingly only the CVB selector correctly identified the frequency band corresponding to the cyclic impulses, based on the distribution of their amplitudes. The kurtosis and Alpha selector indicate both cyclic and non-cyclic impulses frequency ranges. For more details and comparison with other selectors, see [14].

4 Real signal analysis

In this part, we present an application of the statistics we are considering to a real signal. The effectiveness of the different methods is verified on the data from the bearing of a crushing machine; see Figure 2.

However, due to the lack of local faults in the considered vibration data, we introduce an artificial component related to local damage. A similar approach was performed in [13, 44, 47]. The signal is presented in Figure 4. The length of the signal is 6 seconds and the sampling frequency is 25 kHz. The local fault was added with frequency modulation equal to 30 kHz and carrier frequency equal to 2.5 kHz (2–3 kHz). The spectrogram of the data is presented in Figure 11 (a). As we can see, the data reveals high-energy wide-band impulses around 0.25, 4.5 and 5.25 seconds, which correspond to falling rocks. The real vibration signal contains various components with a complex structure, and the component related to the damage is almost imperceptible above the noise. We apply the informative frequency band selectors to the real signal



Figure 2. Crushing machine in a copper ore mine [45]

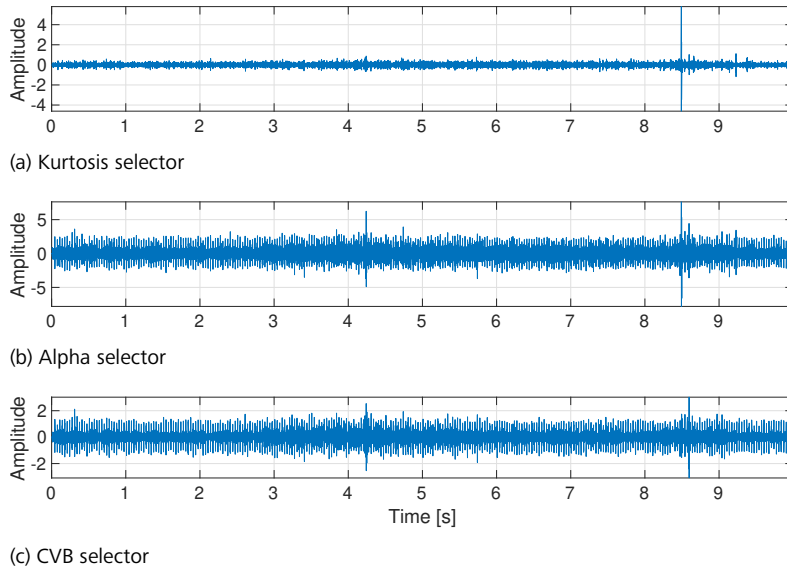


Figure 3. Results of the copper ore crusher's signal filtration performed by three different selectors [13]

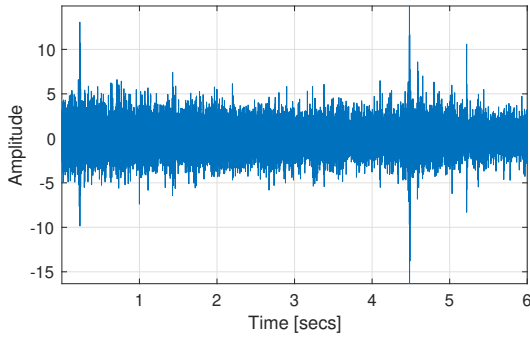


Figure 4. Analysed real signal [14]

with its added fault. The results are presented in Figure 11 (b)–(d). The amplitude of the non-cyclic impulses is smaller than in the simulated data, but one can see that the results of the kurtosis fail. The Alpha selector properly indicates the IFB, but with much less selectivity than the CVB selector.

The signal filtered using the analysed selectors is presented in Figure 3. Filtering driven by the kurtosis selector (panel (a)) provides a single impulsive component that is a non-informative signal. In contrast, the Alpha selector (panel (b)) allows extraction of cyclic impulsive components with 2 random impulses. Finally, results of the CVB selector-based filtering (panel (c)) show a similar effect to the Alpha selector with slightly smaller random impulses. To summarise, Figure 3 shows that the filtration with the kurtosis selector fails, while the results of Alpha and CVB-based methods are acceptable and similar; however, we note that the shape of the CVB selector is better.

5 Cyclostationary analysis for non-Gaussian signals

In this part, we discuss new research related to the cyclostationary analysis for non-Gaussian signals. More precisely, we propose to apply dependency measures expressed by the known correlation coefficients to detect the cyclic behaviour on the time-frequency map. The idea is as follows. First, we represent the signal as the time-frequency map (Figure 1 (a)), then we consider the corresponding sub-signals $S_1, S_2, \dots, S_k$ as the separate time series (Figure 1 (b)), and finally, we apply the dependency measure $m(\cdot, \cdot)$ (see examples below) to the time series extracted from the time-frequency representation of the signal. The general idea of this approach is illustrated in Figure 5.

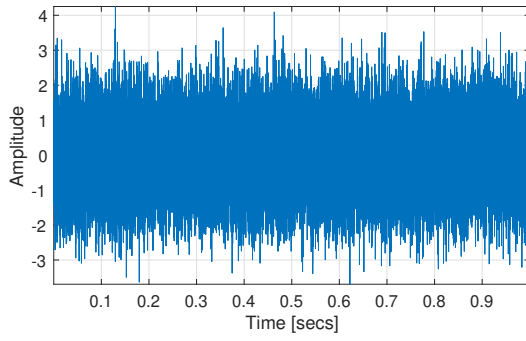
5.1 Pearson correlation map

Let $(x_1, y_1), (x_2, y_2), \dots, (x_N, y_N)$ be a bi-dimensional sample of a random vector (X, Y) , where N is the sample length. The Pearson correlation of (X, Y) is defined as follows [8]:

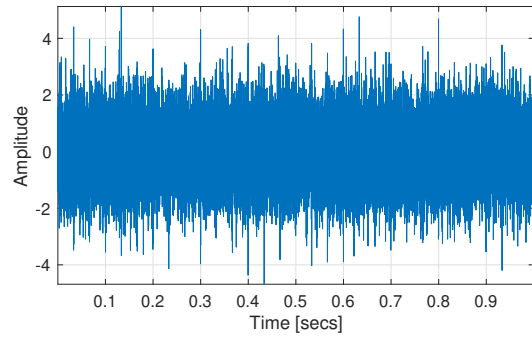
$$\rho_{XY} = \frac{\text{cov}(X, Y)}{\sigma_X \sigma_Y},$$

$$\begin{aligned} &m(s_1, s_1), m(s_1, s_2), \dots, m(s_1, s_l) \\ &m(s_2, s_1), m(s_2, s_2), \dots, m(s_2, s_l) \\ &\vdots \\ &m(s_k, s_1), m(s_k, s_2), \dots, m(s_k, s_l) \end{aligned}$$

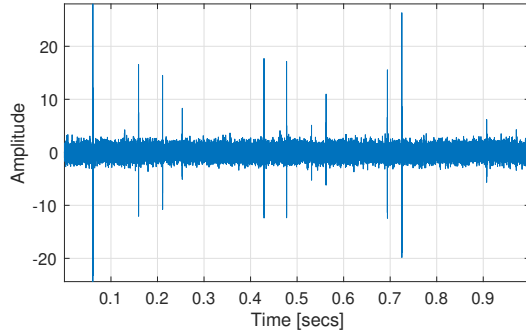
Figure 5. Dependency map structure for spectrogram sub-signals and a given dependency measure $m(\cdot, \cdot)$ [32]



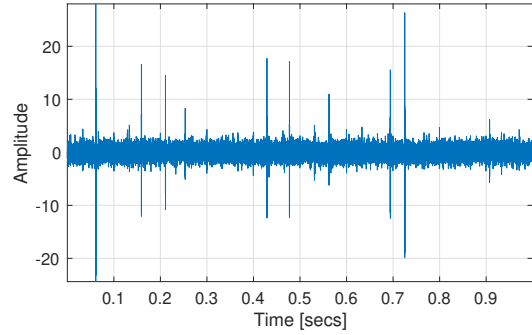
(a) s_1 – Gaussian noise



(b) s_2 – Gaussian noise with cyclic impulses

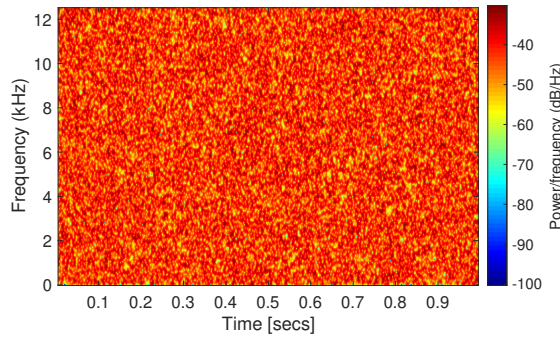


(c) s_3 – non-Gaussian noise

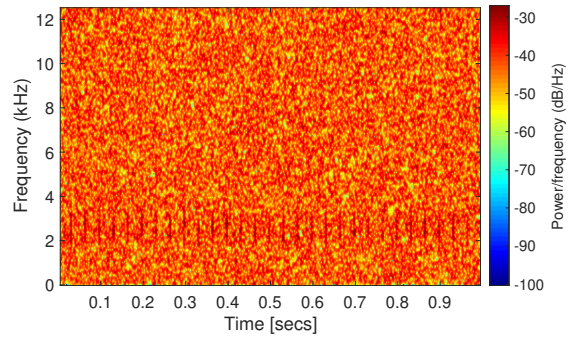


(d) s_4 – non-Gaussian noise with cyclic impulses

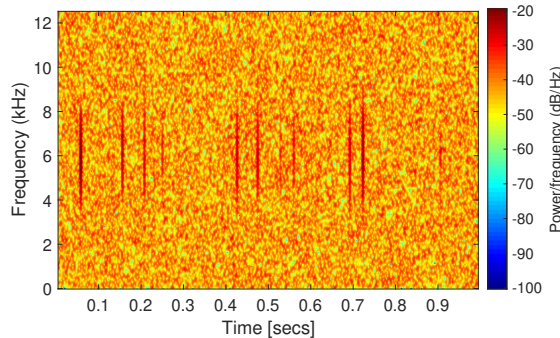
Figure 6. Simulations of the considered signals [14]



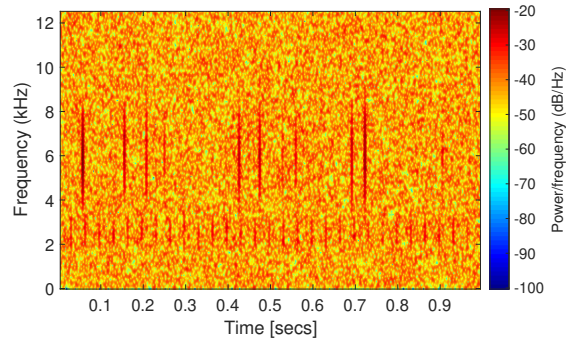
(a) s_1 – Gaussian noise



(b) s_2 – Gaussian noise with cyclic impulses

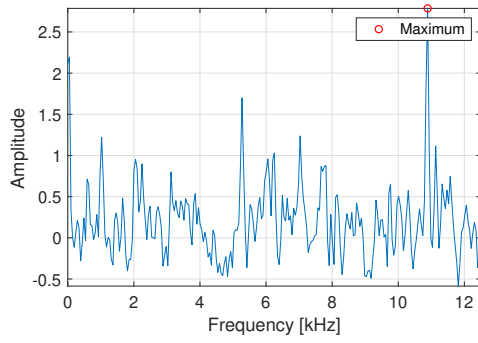


(c) s_3 – non-Gaussian noise

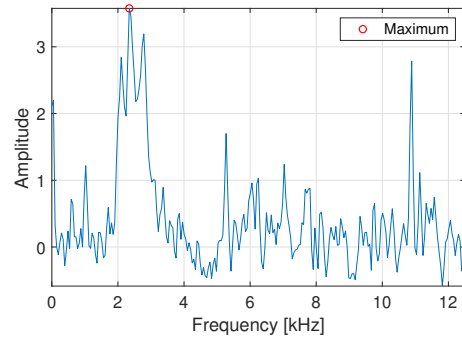


(d) s_4 – non-Gaussian noise with cyclic impulses

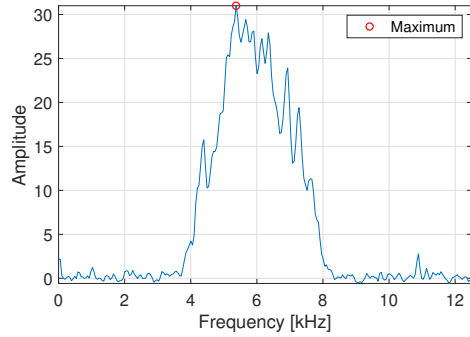
Figure 7. Spectrograms of the simulated signals [14]



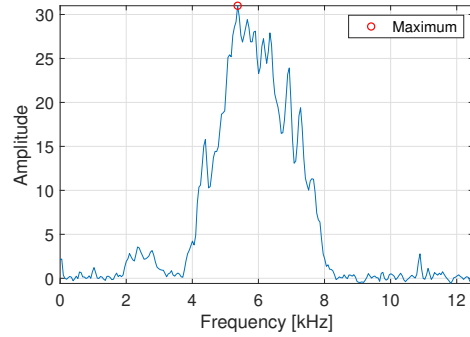
(a) s_1 – Gaussian noise



(b) s_2 – Gaussian noise with cyclic impulses

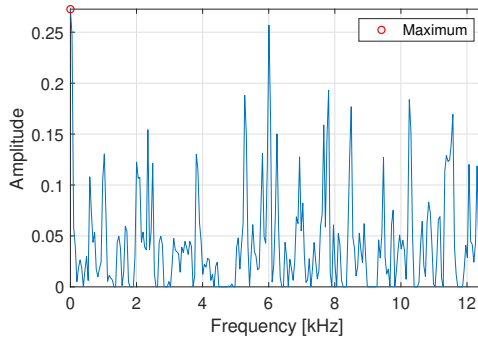


(c) s_3 – non-Gaussian noise

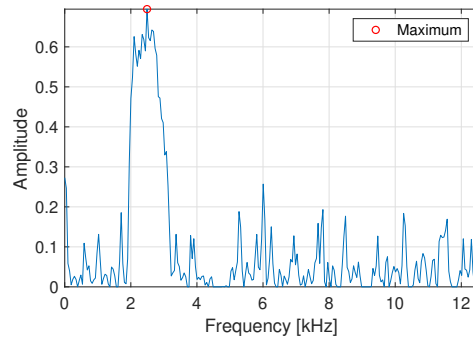


(d) s_4 – non-Gaussian noise with cyclic impulses

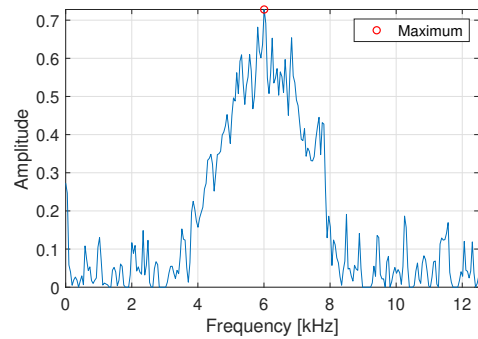
Figure 8. Spectral kurtosis for simulated signals [14]



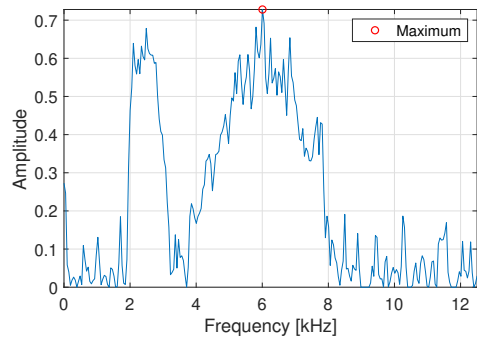
(a) s_1 – Gaussian noise



(b) s_2 – Gaussian noise with cyclic impulses

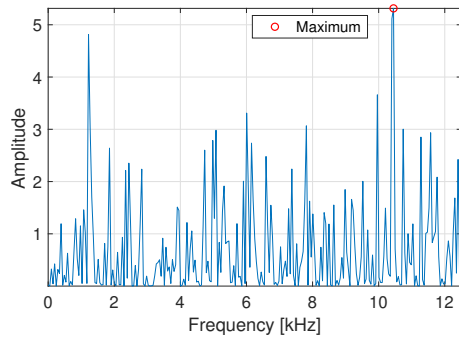


(c) s_3 – non-Gaussian noise

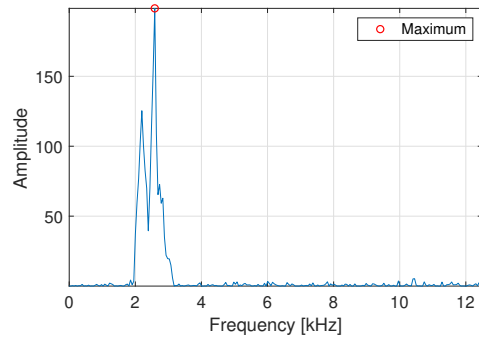


(d) s_4 – non-Gaussian noise with cyclic impulses

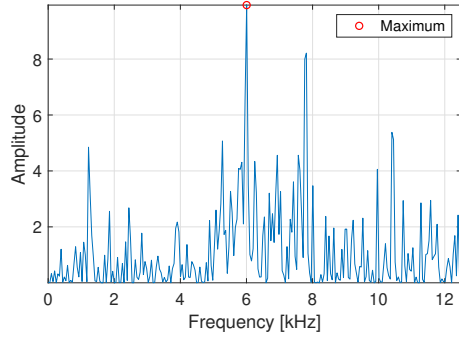
Figure 9. Alpha selector for simulated signals [14]



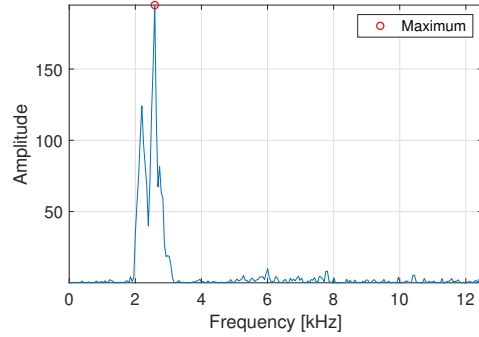
(a) s_1 – Gaussian noise



(b) s_2 – Gaussian noise with cyclic impulses

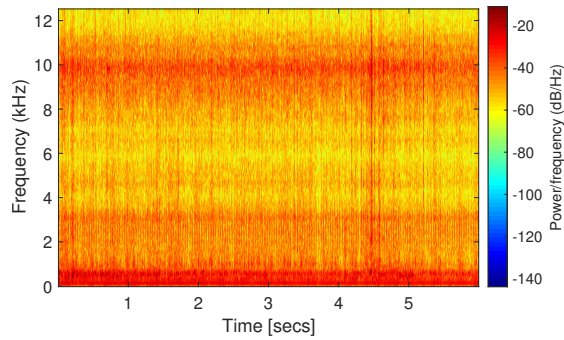


(c) s_3 – non-Gaussian noise

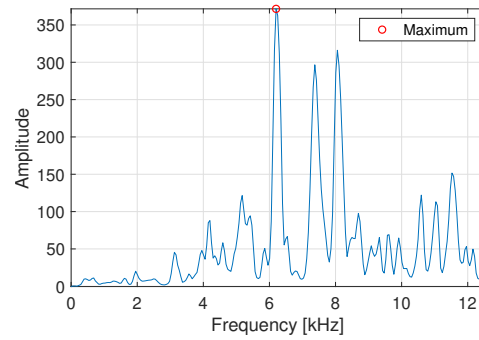


(d) s_4 – non-Gaussian noise with cyclic impulses

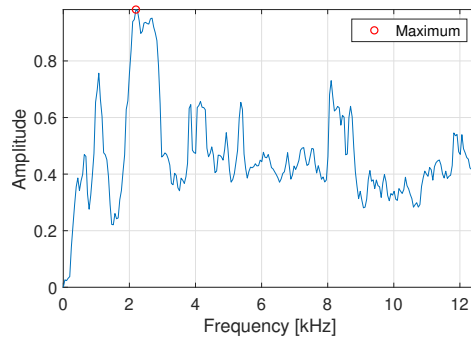
Figure 10. CVB selector for simulated signals [14]



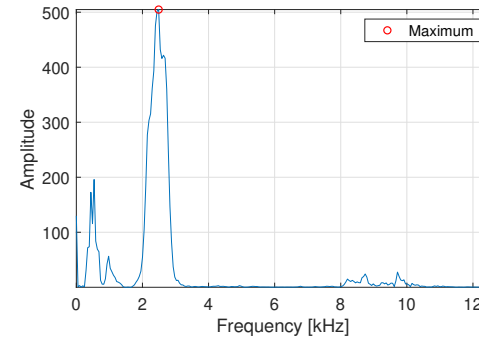
(a) Spectrogram



(b) Kurtosis selector



(c) Alpha selector



(d) CVB selector

Figure 11. Results for real signal from crushing machine [14]

where $\text{cov}(\cdot, \cdot)$ is the covariance function, σ_X is the standard deviation of X , and σ_Y is the standard deviation of Y . The empirical equivalence of ρ_{XY} for $(x_1, y_1), (x_2, y_2), \dots, (x_N, y_N)$, denoted ρ_{xy} , is defined as [8]

$$\rho_{xy} = \frac{\sum_{i=1}^N (x_i - \bar{x})(y_i - \bar{y})}{\sqrt{\sum_{i=1}^N (x_i - \bar{x})^2} \sqrt{\sum_{i=1}^N (y_i - \bar{y})^2}},$$

where $\bar{x}, \bar{y}$ are sample means of data vectors x and y respectively. The Pearson correlation is usually a good dependency measure for finite-variance signals. However, this measure is sensitive to outliers. In our study, the Pearson correlation coefficient is applied to the sub-signals $S_1, S_2, \dots, S_k$; see Figure 1 (b).

5.2 Spearman correlation map

Let us again consider a bi-dimensional sample $(x_1, y_1), (x_2, y_2), \dots, (x_N, y_N)$ corresponding to a random vector (X, Y) . Each pair (X, Y) and its empirical counterpart (x_i, y_i) correspond to a pair (Q, W) and (q_i, w_i) , where q_i is the rank of the observation x_i in the sample $x_1, x_2, \dots, x_N$ and w_i is the rank of the observation y_i in the sample $y_1, y_2, \dots, y_N$. The Spearman rank correlation coefficient for the vector (X, Y) is defined as [23]

$$r_{XY} = \frac{\text{cov}(Q, W)}{\sigma_Q \sigma_W},$$

where $\text{cov}(\cdot, \cdot)$ is the covariance function, σ_Q and σ_W are the standard deviations of the rank variables. The empirical version of the Spearman correlation coefficient is given by

$$r_{xy} = \frac{\frac{1}{N-1} \sum_{i=1}^N (q_i - \bar{q})(w_i - \bar{w})}{\left[\frac{1}{N-1} \sum_{i=1}^N (q_i - \bar{q})^2 \frac{1}{N-1} \sum_{i=1}^N (w_i - \bar{w})^2 \right]^{1/2}},$$

where $\bar{q}$ and $\bar{w}$ are sample means in the relevant rank samples.

The Spearman correlation takes values in the interval $[-1, 1]$ and investigates a monotonic relationship, in contrast to the Pearson correlation, which analyses a linear relationship. Outliers do not disturb the Spearman correlation, whereas they do in the case of the Pearson correlation.

As we did for the Pearson correlation, in our study, we apply the Spearman correlation to the sub-signals $S_1, S_2, \dots, S_k$; see Figure 1 (b).

5.3 Kendall correlation map

The formula for the Kendall τ coefficient can be written as follows [22]:

$$\tau = \frac{2}{N(N-1)} \sum_{1 \leq i < j \leq N} J((x_i, y_i), (x_j, y_j)),$$

where

$$J((x_i, y_i), (x_j, y_j)) = \text{sgn}(x_i - x_j) \text{sgn}(y_i - y_j)$$

and $J((x_i, y_i), (x_j, y_j)) = 1$ if a pair (x_i, y_i) is concordant with a pair (x_j, y_j) , i.e. if $(x_i - x_j)(y_i - y_j) > 0$; $J((x_i, y_i), (x_j, y_j)) = -1$ if a pair (x_i, y_i) is discordant with a pair (x_j, y_j) , i.e. if $(x_i - x_j)(y_i - y_j) < 0$.

The Kendall correlation coefficient is based on the difference between the probability that two variables are in the same order (for the observed data vector) and the probability that their order is different. The Kendall coefficient indicates not only the strength but also the direction of the dependency. Like the Spearman correlation, it is resistant to outliers. In our study, the Kendall correlation is applied to the sub-signals from the time-frequency representation of the signal.

In the study [32], additional enhancements of the proposed dependency maps were proposed. We refer the readers to this position for more details.

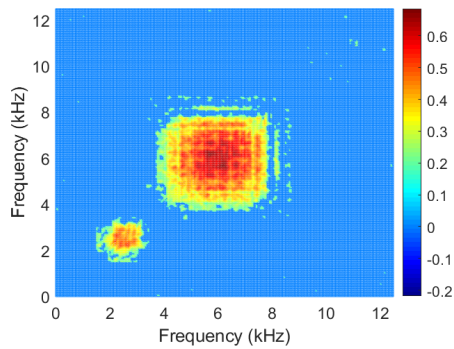
6 Comparative study of dependency measure applications – analysis of simulated and real signals

In this section, we present results of IFB selection by using the above-mentioned dependency measures. The analyses are performed for the simulated signal s_4 , presented in Figure 6 (d). Recall that this is the signal that contains the non-Gaussian noise with cyclic impulses. In Figure 12, we present the enhanced dependency maps for the three correlation coefficients and for the signal s_4 by using the procedure presented in [32].

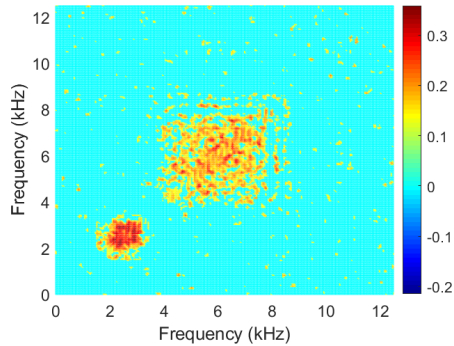
As one can see in Figure 12, in the case of the Pearson correlation map, the result differs from the other correlation maps. The correlation values at the location of the non-cyclic impulses are higher than for the cyclic impulses. In the other cases, one can see the opposite result. This result indicates that the application of more robust dependency measures may help to identify the cyclic impulsive behaviour in the case of impulsive noise.

Finally, we apply the dependency measures to the real signal from the crushing machine presented in Section 4. The real signal is depicted in Figure 4. In Figure 13, we present the enhanced dependency maps based on the Pearson, Spearman and Kendall correlations discussed in the previous section. We can see that the clear picture can be obtained using the robust measures (i.e. Spearman and Kendall correlation maps), where the IFB is indicated properly. The Pearson correlation map reacts to the frequency band related to the non-cyclic impulses; thus it cannot be used as a proper measure for the cyclostationary analysis in the case of non-Gaussian noise. Related discussions are presented in [25].

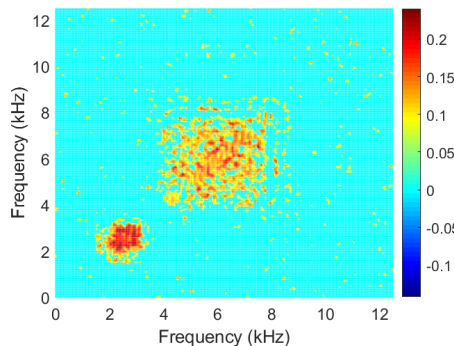
Acknowledgements. Let me thank Krzysztof Burnecki for the kind invitation to write this article. I would also like to thank Radosław Zimroz and the scientists in the Digital Mining Center (Faculty of Geoengineering, Mining and Geology at Wrocław University of Science and Technology, WUST) for our almost ten years of



(a) Pearson correlation

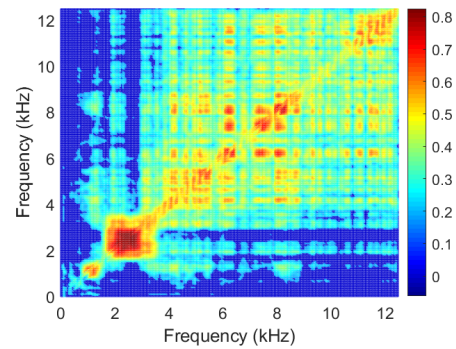


(b) Spearman correlation

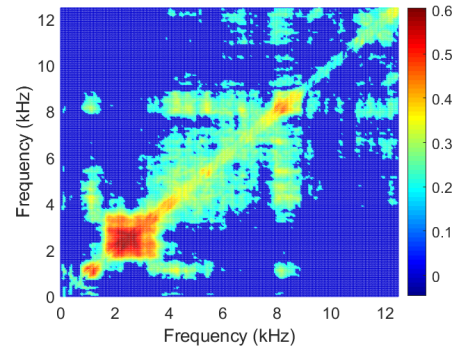


(c) Kendall correlation

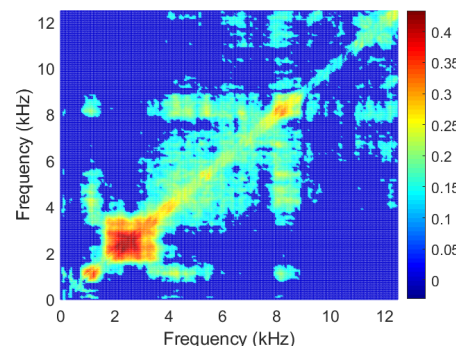
Figure 12. Enhanced correlation maps for the simulated signal s_4 [32]



(a) Pearson correlation



(b) Spearman correlation



(c) Kendall correlation

Figure 13. Enhanced correlation maps for the real signal s_4 [32]

interdisciplinary research on new methods for real signal analysis. Thank you for showing me that mathematics can really be useful in practical applications, e.g. in the mining industry. Finally, I thank my PhD students and indeed all my students for providing new insights into our research, and positive energy.

The research was done within the Hugo Steinhaus Center at WUST whose goal is to organise, encourage and support research and education in numerical and stochastic techniques as applied in science and technology.

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